Optimal Control of Wave Energy Converters: From Adaptive PI Control to Model Predictive Control

Hoai-Nam Nguyen

Electronics and Physics Department - Telecom SudParis

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Outline



- Problem Formulation
- 3 Adaptive PI Control
- 4 Model Predictive Control
- 5 Experimental Results

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- Problem Formulation
- 3 Adaptive PI Control
- 4 Model Predictive Control
- 5 Experimental Results

Short Bio (1/3)

- Electrical Engineer, 09/2002-06/2009, Bauman Moscow State Technical University.
- Ph.D. in Automatic Control, 10/2009-09/2012, Supélec, Gif-sur-Yvette.



Short Bio (2/3)

- Post-doc, 10/2012-09/2014, Technion Israel Institute of Technology, Haifa.
- Research Engineer, 10/2014-09/2021, IFP Energies Nouvelles, Lyon.





Short Bio (3/3)

- Since 09/2021, Associate Professor, Electronics and Physics Department, Telecom-SudParis.
- Research Interests: Sensor Fusion, Kalman/Bayesian Estimation, Signal Processing, Constrained Optimal Control, Real time Optimization.



Outline

Short Bio

- Problem Formulation
 - Wave Energy Converter
 - WEC Modeling
 - Control Objective
- 3 Adaptive PI Control
- 4 Model Predictive Control

Experimental Results

Wave Energy Converter

- WEC is a device which captures the power of waves and transforms it to electricity.
- The electricity generation from waves could amount to more than 2TW (18000TWh/year)



Wave energy distribution kW/m

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Some Prototypes



Some Prototypes, cont.



WEC modeling



• Newton's second law for rotation

 $J\ddot{ heta}(t) = M_{ex}(t) - M_{PTO}(t) - M_{hd}(t) - M_{rad}(t)$

• $\theta(t)$: Float angle

WEC modeling, cont.

- J : Mass inertia moment
- $M_{ex}(t)$: Wave excitation moment
- $M_{PTO}(t)$: PTO moment
- $M_{hd}(t)$: Hydrostatic moment (due to gravity)

 $M_{hd}(t) = K\theta(t)$

• $M_{rad}(t)$: Radiation moment (due to the float movement)

$$M_{rad}(t) = \int_{ au=0}^{t} h(t- au) \dot{ heta}(au) d(au)$$

WEC modeling, cont.

- J, K, h(t) can be derived from boundary element methods, estimated via dedicated experiments or both.
- State space equation,

$$\begin{cases} \dot{x}(t) = Ax(t) + B(M_{ex}(t) - M_{PTO}(t)) \\ y(t) = Cx(t) \end{cases}$$

Where

•
$$x(t)$$
: State
• $y(t) = \begin{bmatrix} \theta(t) \\ \dot{\theta}(t) \end{bmatrix}$: Output

Control Objective

• Find $M_{PTO}(t)$ to maximize

$$P_{avr} = rac{1}{T} \int_0^T \mu(t) \dot{ heta}(t) M_{PTO}(t) dt$$

• $\mu(t)$: efficiency coefficient, that depends on $\dot{\theta}(t)M_{PTO}(t)$

$$\mu = \left\{ \begin{array}{ll} 0.7, & \text{if } \dot{\theta} M_{\text{PTO}} \geq 0 \\ \frac{1}{0.7} = 1.43, & \text{if } \dot{\theta} M_{\text{PTO}} < 0 \end{array} \right.$$

• Pay two times more expensive to use energy from network

Control Objective, cont.

Problems

- **1** Even the control system is linear, the cost function is nonlinear.
- 2 x(t), $M_{ex}(t)$ are not directly available.
- Input and state constraints.

Solutions

- Adaptive PI control.
- Model predictive control.

Outline



2 Problem Formulation

Adaptive PI Control
PI control for regular waves
Adaptive PI control

Model Predictive Control

Experimental Results

PI control for regular waves

• Assumption: $M_{ex}(t)$ is available and

 $M_{ex}(t) = A_w sin(wt + \phi)$

• WEC model in the frequency domain

$$\frac{v(jw)}{M_{ex}(jw) - M_{PTO}(jw)} = \frac{1}{Z(jw)}$$

- Where
 - $v(jw) = \dot{\theta}(jw)$
 - Z(jw) : intrinsic impedance

PI control for regular waves, cont.

• Idea: Design a linear control law

 $M_{PTO}(jw) = K(jw)v(jw)$

that maximizes the cost

Denote

$$K(jw) = R_k + jX_k, \quad Z(jw) = R_z + jX_z$$

Theorem

$$P_{avr} = \frac{A_w^2 \left(\mu R_k + \frac{1}{\pi} (\mu - \frac{1}{\mu}) R_k (\frac{X_k}{R_k} - \arctan(\frac{X_k}{R_k})) \right)}{2((X_z + X_k)^2 + (R_z + R_k)^2)}$$

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Result of Falness et al., with $\mu = 1$

• With $\mu = 1$, recover the well-known result of Falness et al.

$$P_{avr} = \frac{A_w^2 R_k}{2((X_z + X_k)^2 + (R_z + R_k)^2)}$$

• Pavr is maximal, iff

$$X_k = -X_z, \ R_k = R_z$$

Hence

$$\frac{v(jw)}{M_{ex}(jw)} = \frac{1}{2R_z(w)}$$

• P_{avr} is maximal iff v(t) is in phase with $M_{ex}(t)$

• Are the results of Falness et al. correct also for $\mu < 1$?

Optimal Control of WEC

Falness et al., cont.

- Let's take $X_k = -X_z$ and $R_k = R_z$ for P_{avr} with $\mu = 0.7$
- *P_{avr}* < 0 for all ω ≤ 5.5(*rad/sec*) (where the wave has the most energy)
- Solution is not optimal, since one can take $R_k = X_k = 0$



PI control for regular waves, cont.



- It can be shown that the cost function is convex. Hence the optimal solution is unique.
- v(t) is generally not in phase with $M_{ex}(t)$, since $X_k \neq -X_z$.

PI control for regular waves, cont.

• Up to now, for each regular wave $M_{ex}(t) = A_w sin(wt + \phi)$, the frequency response of the optimal controller is calculated

 $K(jw) = R_k(w) + jX_k(w)$

• If K(jw) is chosen as a PI controller

$$K(jw) = K_p + rac{K_i}{jw}$$

Then

$$K_p = K_k(w), \ K_i = -wX_k(w)$$

Problems

- $M_{ex}(t)$ is not measurable.
- 2 Real $M_{ex}(t)$ is not a sinusoid.



Only the second problem is addressed now.

Frequency estimation

- Real $M_{ex}(t)$ is not a sinusoid, but it is not far from the sinusoid.
- Idea: approximate ON-LINE $M_{ex}(t)$ as

 $M_{ex}(t) = A_w(t)sin(w(t)t + \phi(t))$

where $A_w(t), w(t), \phi(t)$ are parameters.

• Classical problem.

Frequency estimation, cont.

- Unscented Kalman filter is used to estimate $A_w(t), w(t), \phi(t)$
- Details are not presented here.



Frequency estimation, cont.



Adaptive PI control

The adaptive PI control algorithm is summarized as follows,

- 1. Measure p(t), v(t)
- 2. Estimate $M_{ex}(t)$, $A_w(t), w(t), \phi(t)$

 $M_{ex}(t) = A_w(t)sin(w(t)t+\phi(t))$



3. Calculate the control action

$$M_{PTO}(t) = \mathcal{K}_{p}(w) v(t) + \mathcal{K}_{i}(w) \int_{0}^{t} v(au) d au$$

Outline



2 Problem Formulation

3 Adaptive PI Control

Model Predictive Control

- Model prediction control Basic concepts
- Wave excitation moment estimation
- Wave excitation moment prediction
- Weighted MPC

Experimental Results

Model predictive control - Basic concepts



A model of the process is used to predict the future evolution of the process to optimize the control signal

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Receding horizon philosophy

At time k: Solve an optimal control problem over a finite future horizon of N steps

 $\min_{\substack{u(k), u(k+1), \dots, u(k+N-1)}} J(k),$ s.t. $\begin{cases} u_{min} \le u(k+j) \le u_{max}, \\ y_{min} \le y(k+j) \le y_{max} \end{cases}$

J(k) : cost function



- Only apply the first move u^{*}(k)
- At time k + 1: Get new measurements, repeat the optimization. And so on ...

Why MPC for WEC?

- Maximize the extracted energy.
- Input and state constraints are incorporated in the design phase.
- Nonlinear efficiency coefficient is considered in the design phase.
- Wave prediction is explicitly used.

MPC problems

- Wave excitation moment at the present and in the future are required
 - Wave moment estimation
 - Wave prediction
- Nonlinear and non-convex optimization problem due to the nonlinear efficiency coefficient

Wave excitation moment estimation

- Idea: Use a WEC model + measured outputs to estimate $M_{ex}(t)$
- Not a new idea
 - $M_{ex}(t)$ is decomposed as, P. Kracht et al., 2014

$$M_{ex}(t) = \sum_{j=1}^{m} \alpha_j(t) sin(\omega_j t) + \beta_j(t) cos(\omega_j t)$$

where ω_i are chosen

• $\alpha_j(t)$, $\beta_j(t)$ are estimated online using Luenberger observer

Wave excitation moment estimation, cont.

- The choice of ω_i is crucial
- The approach was experimentally tested at the Aalborg university
 - Slightly overestimate the amplitude
 - Non-negligible delay
- Not reliable in practice, since ω_j are time-varying



Random walk approach

• Idea: see $M_{ex}(k)$ as a state

 $M_{ex}(k+1) = M_{ex}(k) + \epsilon(k)$

 $\epsilon(k)$: variation of $M_{ex}(k)$, and is considered as a noise

Hence

$$\begin{cases} \begin{bmatrix} x \\ M_{ex} \end{bmatrix}^{+} = \begin{bmatrix} A & B \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x \\ M_{ex} \end{bmatrix} - \begin{bmatrix} B \\ 0 \end{bmatrix} M_{PTO} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \epsilon$$
$$y = \begin{bmatrix} C & D \end{bmatrix} \begin{bmatrix} x \\ M_{ex} \end{bmatrix} - DM_{PTO}$$

Random walk approach, cont.

- The problem of estimating $M_{ex}(k)$ becomes the state estimation problem
- Kalman filter is used for this purpose
- Clearly, the approach can be used to estimate any kind of M_{ex} (not necessarily periodic)

Experimental results



Wave excitation moment prediction

- Given wave moments y(k), $k = 0, 1, ..., k_0$ until time k_0
- Predict wave moments at times $k_0 + 1, k_0 + 2, \dots, k_0 + N$



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AR model based forecast

 Idea: Wave moment at time k is a linear function of a number p of its past values

 $y(k+1) = a_1y(k) + a_2y(k-1) + \ldots + a_py(k-p+1)$

- $\{a_1, a_2, \ldots, a_p\}$: parameters
- {*a*₁, *a*₂..., *a_p*} can be found by minimizing the one step ahead prediction error

$$\min_{a_1,a_2,...,a_p} \sum_{j=p+1}^k (y(j) - \sum_{i=1}^p a_i y(j-i))^2$$

AR model based forecast, cont.

- Least square problem. Solution can be found analytically
- Result is not satisfactory for prediction
- Fusco's and Ringwood's idea: Long Rang Predictive Identification, i.e. minimizing not only the one step, but also the two-step, ..., the h-step prediction errors
- Nonlinear least square optimization problem. Batch-processing based solution.

Filter bank based forecast

- Previous method: iterative forecast
- Idea: forecast each horizon independently from the others
- N models for forecasting N steps ahead

$$\begin{cases} y(k+1) = a_{11}y(k) + a_{12}y(k-1) + \dots + a_{1p}y(k-p+1) \\ y(k+2) = a_{21}y(k) + a_{22}y(k-1) + \dots + a_{2p}y(k-p+1) \\ \vdots \\ y(k+N) = a_{N1}y(k) + a_{N2}y(k-1) + \dots + a_{Np}y(k-p+1) \end{cases}$$

• Unknown parameters a_{ij} , $i = \overline{1, N}$, $j = \overline{1, p}$ are estimated by Kalman filter

Filter bank based forecast

- To forecast N steps ahead, one needs N Kalman filters
- Computational complexity is higher than iterative forecast
- Performance is better

Nonlinear MPC

- Finally, we are getting to the point
- Recall the state space equation of WEC

$$\begin{cases} x(k+1) = Ax(k) + BM_{ex}(k) - Bu(k) \\ y(k) = Cx(k) + DM_{ex}(k) - Du(k) \end{cases}$$

where $u(k) = M_{PTO}(k)$.

• Cost function at time k,

$$\max_{u(k),...,u(k+N)} \sum_{j=0}^{j=N} \mu(k+j)v(k+j)u(k+j)$$

 μ : nonlinear efficiency coefficient

Nonlinear efficiency function

• Taking into account directly μ in the cost gives rise to a nonlinear and nonconvex optimization problem

$$\min_{U(k)} (U(k)^T H(\mu) U(k) + f(\mu)^T U(k))$$

s.t. $u_{min} \le u(k+j) \le u_{max}, j = \overline{0, N}$

- Issues
 - Computational load
 - Difficult to investigate: feasibility, stability, robustness



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Weighted MPC

• Consider again the cost function, for $\mu=1$

 $J = \max_{U(k)} (u(k)v(k) + u(k)v(k+1) + u(k+1)v(k+1) + \ldots)$

- Weights are equal for all future costs
- This is not reasonable, since
 - Wave prediction performance is better for a short horizon, than for a large horizon.
 - It is better to put high weights on the current obtained energy for the first few time instants.

Weighted MPC, cont.

In the result

 $J = \max_{U(k)} (w_0 u(k) v(k) + w_0 u(k) v(k+1) + w_1 u(k+1) v(k+1) + \ldots)$

- $w_0, w_1, \ldots, w_{N-1}$: tunning coefficients
- We usually choose

 $w_0 \geq w_1 \geq \ldots \geq w_{N-1}$

• QP problem

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Experimental Setup



- Tests in Aalborg University basin on a pivoting-buoy point absorber.
- 4 different sea states are considered.



Experimental results, cont



Conclusions

- Two main solutions are proposed for WECs with non-perfect PTO
 - Adaptive PI: optimal control for regular waves, wave force and dominant wave frequency estimation.
 - Model predictive control: wave estimation, wave prediction, weighted MPC, QP problem.
- Successfully implemented for a real system.
- Perspective: Decentralized/distributed control, stochastic MPC.

THANK YOU